

MODULE 6

DISCRETE TIME FOURIER TRANSFORM

- Discrete time fourier transform of a signal $x(n)$ is defined as

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x(n) \cdot e^{j\omega n}$$

- Discrete time FT can be applied only for discrete time signals

- Continuous time FT of a signal $x(t)$ is

$$X(\omega) = \int_{-\infty}^{\infty} x(t) \cdot e^{-j\omega t} \cdot dt$$

- The existence condition of CTFT

is $\int_{-\infty}^{\infty} x(t) \cdot dt < \infty$

- The existence condition of DTFT

is $\sum_{n=-\infty}^{\infty} x(n) < \infty$

- Discrete time Fourier transform & Z transform are interrelated

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Page _____

Relation b/w z transform and Discrete time Fourier transform

The z transform of a signal $x(n]$ is

$$X(z) = \sum_{n=-\infty}^{\infty} x(n) \cdot z^{-n}$$

by putting $z = e^{j\omega}$, the above expression changed in to DTFT

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x(n) \cdot (e^{j\omega})^{-n}$$

$$= \sum_{n=-\infty}^{\infty} x(n) \cdot e^{-j\omega n}$$

i.e. DTFT of $x(n) = z$ transform of $x(n) |_{z=e^{j\omega}}$

$$X(e^{j\omega}) = X(z) |_{z=e^{j\omega}}$$

This is the Relation between z transform
and Discrete time Fourier transform:

Simple method to calculate DTFT

1. If a signal $x(n]$ is given, and the existence criteria of signal is satisfied. Then we find the z transform of that signal. After that we put $z = e^{j\omega}$ to get its DTFT

Q. Find the DTFT of $x(n) = 2^n \cdot u(n)$.

Existence criteria is not satisfied. This is an unbounded signal ($a > 1$) so, DTFT of that signal doesn't exist

Q. Find the DTFT of $x(n) = (\frac{1}{2})^n \cdot u(n)$

Existence condn is satisfied

$$X(z) = \frac{1}{1 - \frac{1}{2}z^{-1}}$$

$$z = e^{j\omega}$$

$$X(e^{j\omega}) = \frac{1}{1 - \frac{1}{2}e^{-j\omega}}$$

Let $\omega = \omega + 2\pi$

$$X(e^{j(\omega + 2\pi)}) = \frac{1}{1 - \frac{1}{2} e^{-j(\omega + 2\pi)}}$$

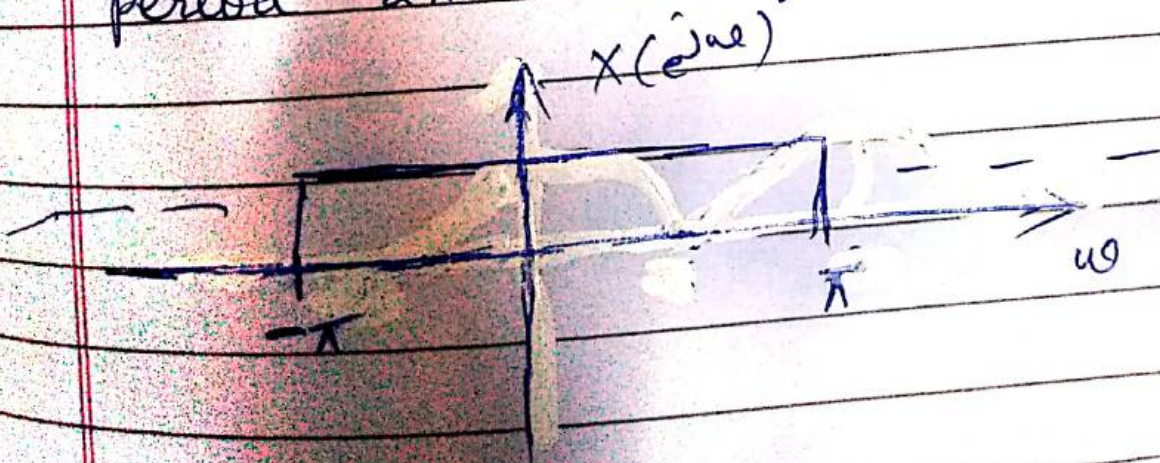
$$= \frac{1}{1 - \frac{1}{2} e^{-j\omega} \cdot e^{-j2\pi}}$$

$$= \frac{1}{1 - \frac{1}{2} e^{-j\omega}} = X(e^{j\omega})$$

$$X(e^{j(\omega + 2\pi)}) = X(e^{j\omega})$$

$X(e^{j\omega})$ is periodic with period of 2π .

In the DTFT, signal $x(n)$ is discrete in time domain, and its transform is periodic in frequency domain with period 2π . Its frequency spectrum is continuous also.



The inverse FT of discrete time signal.

$$x(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) \cdot e^{j\omega n} d\omega$$

The inverse FT of CT signal is

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega) \cdot e^{j\omega t} d\omega$$

Here the frequency domain is not periodic, is why we are taking $-\infty$ to ∞ ;

Q. Find the DTFT of the following signals

1) $x(n] = \delta(n)$.

Z transform $X(z) = 1$

$z = e^{j\omega}$ $X(e^{j\omega}) = 1$

2) $x(n] = u(n)$.

$X(z) = \sum_{n=0}^{\infty} 1 \cdot z^{-n}$

$= \frac{1}{1-z^{-1}}$

$z = e^{j\omega}$

$X(e^{j\omega}) = \frac{1}{1-e^{-j\omega}}$

3) $x(n] = u(n-k)$

$X(z) = \sum_{n=k}^{\infty} 1 \cdot z^{-n}$

$= \frac{z^{-k}}{z} + \frac{z^{-(k+1)}}{z} + \frac{z^{-(k+2)}}{z} + \dots$

$= z^{-k} [1 + z^{-1} + z^{-2} + \dots]$

$= z^{-k} \left[\frac{1}{1-z^{-1}} \right]$

$z = e^{j\omega}$

$X(e^{j\omega}) = e^{-j\omega k} \frac{1}{1-e^{-j\omega}}$

4. $x(n) = a^n \cdot u(n)$

$$X(z) = \frac{1}{1 - az^{-1}}$$

$$z = e^{j\omega}$$

$$X(e^{j\omega}) = \frac{1}{1 - ae^{-j\omega}}$$

5. $x(n) = -a^n u(-n-1)$

$$X(z) = \frac{1}{1 - az^{-1}}$$

$$= \frac{1}{1 - ae^{j\omega}}$$

6. $x(n) = (-a)^n \cdot u(n)$

$$X(z) = \frac{1}{1 + az^{-1}}$$

$$z = e^{j\omega}$$

$$X(e^{j\omega}) = \frac{1}{1 + ae^{-j\omega}}$$

7:

$$7. x(n) = \delta(n-k)$$

$$X(z) = \sum_{n=-\infty}^{\infty} x(n) \cdot z^{-n}$$

$$= \sum_k \delta(n-k) \cdot z^{-n}$$

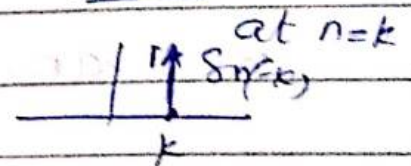
$$z = e^{j\omega} \quad \Rightarrow \quad z^{-k}$$

$$X(e^{j\omega}) = \underline{e^{-j\omega k}}$$

$$\begin{aligned} n &= k \\ \delta(n-k) &= 0 \\ z^{-n} &= \underline{z^{-k}} \end{aligned}$$

Shifted value of impulse multiplied by a signal equal to that signal with 'n' replaced with the existence value of $\delta(n)$

Here $\underline{\delta(n-k)} = 1$



$$8. x(n) = \delta(n+a) - \delta(n-a)$$

$$\begin{aligned} X(z) &= e^{j\omega a} - e^{-j\omega a} \\ &= z^a (1) - z^{-a} (1) \end{aligned}$$

$$\text{put } z = e^{j\omega}$$

$$X(e^{j\omega}) = e^{j\omega a} - e^{-j\omega a}$$

$$= \underline{2j \sin \omega a}$$

$$\text{SO: } \delta(n-k) \cdot z^{-n} \\ \underline{z^{-k}}$$

use the shifting property of z

$$z(x(n+m)) = z^m \cdot X(z)$$

$$z(x(n-m)) = z^{-m} \cdot X(z)$$

9. Find the DTFT of the following signals.

1. $x(n] = \{1, -1, 2, 2\}$

$$X(z) = 1 - z^{-1} + 2z^{-2} + 2z^{-3}$$

$$z = e^{j\omega}$$

$$X(e^{j\omega}) = 1 - e^{-j\omega} + 2e^{-2j\omega} + 2e^{-3j\omega}$$

2. $x(n] = 2^n \cdot u(n]$

$2^n u(n]$ is not absolutely summable

$a > 1 \Rightarrow \text{exp rising signal}$. so we can't find its DTFT!

$$x(n) = (-0.5)^n u(n) + 2^n \cdot (-n-1) \cdot u(-n-1)$$

$$X(z) = \sum_{n=0}^{\infty} (-0.5)^n z^{-n} + \sum_{n=-\infty}^{-1} 2^n \cdot z^{-n}$$

$$X(z) = \sum_{n=0}^{\infty} (0.5 z^{-1})^n + \sum_{n=1}^{\infty} \frac{1}{2} z^n$$

$$= \sum_{n=0}^{\infty} (0.5 z^{-1})^n + \sum_{n=1}^{\infty} (z^{-1} z)^n$$

$$= \frac{1}{1 - 0.5 z^{-1}} + \frac{1}{2z} \left[\frac{1}{1 - z^{-1} z} \right]$$

Put $z = e^{j\omega}$

$$X(e^{j\omega}) = \frac{1}{1 - 0.5 e^{-j\omega}} + \frac{0.5 e^{j\omega}}{1 - 0.5 e^{j\omega}}$$

$$= \frac{1 - 0.5 e^{j\omega} + 0.5 e^{-j\omega} - 0.25 e^{-j\omega} e^{j\omega}}{1 - 0.5 e^{j\omega} - 0.5 e^{-j\omega} + 0.25}$$

$$= \frac{1 - 0.25}{1 - 0.5(e^{j\omega} + e^{-j\omega}) + 0.25}$$

$$= \frac{0.75}{1 - 2 \times 0.5 \cos \omega + 0.25} = \frac{0.75}{1.25 - \cos \omega}$$

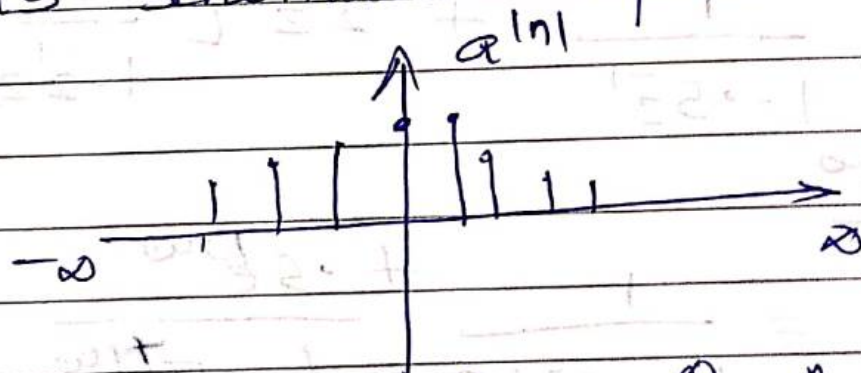
Q. Find the Fourier transform of the signal

$$x(n) = a^{|n|}$$

$a^{|n|}$ is defined as

$$x(n) = \begin{cases} a^n & \text{for } n > 0 \\ \bar{a}^n & \text{for } n < 0. \end{cases}$$

It's schematic representation is



$$X(z) = \sum_{n=-\infty}^{\infty} \bar{a}^n \cdot z^{-n} + \sum_{n=0}^{\infty} a^n \cdot z^{-n}$$

$$= \sum_{n=1}^{\infty} a^n z^n + \sum_{n=0}^{\infty} (a z^{-1})^n$$

$$= a z \left[\frac{1}{1 - a z} \right] + \frac{1}{1 - a z^{-1}}$$

Put $z = e^{j\omega}$

$$X(e^{j\omega}) = a e^{j\omega} \cdot \frac{1}{1 - a e^{j\omega}} + \frac{1}{1 - a e^{-j\omega}}$$

on cross multiplying, we get

$$\frac{ae^{j\omega}(1 - ae^{-j\omega}) + 1 - ae^{j\omega}}{(1 - ae^{j\omega})(1 - ae^{-j\omega})}$$

$$= \frac{ae^{j\omega} - a^2 + 1 - ae^{j\omega}}{(1 - ae^{j\omega})(1 - ae^{-j\omega})}$$

$$1 - ae^{-j\omega} - ae^{j\omega} + a^2$$

$$= \frac{1 - a^2}{1 - 2a \cos \omega + a^2}$$

$$1 - 2a \cos \omega + a^2$$